Topic 3 - Part 1: LTI systems in transformed domains

Linear systems and circuit applications

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Based also on Professor Óscar Barquero Perez, Andrés Martínez and José Luis Rojo's slides

- Why do we "pass" to another domain different from the time domain?
- the answer in the next slides...

- Many signals in nature and communications can be expressed as linear combinations of real/complex exponentials.
- For periodic signals, this will be completely clear after looking the Fourier series.

$$e^{st} - - \exp(st)$$

Note the position of "s" and "t" can be switched:

$$t \in \mathbb{R} - -s \in ?$$

• "s" is a complex number:

$$s \in \mathbb{C}$$

$$s = \sigma + j\omega$$

$$s=\sigma+j\omega$$
• Frequency

$$e^{st}=e^{(\sigma+j\omega)t}$$

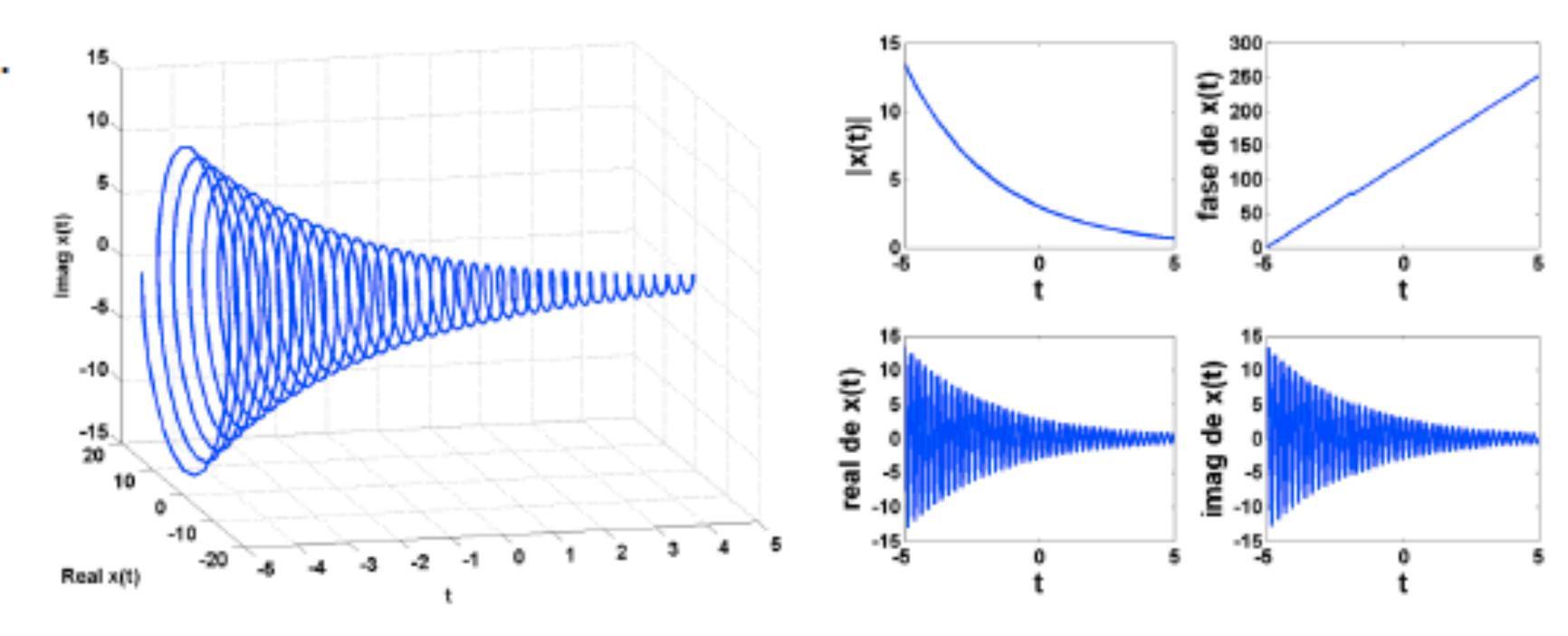
$$=e^{\sigma t+j\omega t}$$

$$=e^{\sigma t}e^{j\omega t}$$
• envelope (real exponential)

• From Laura Martínez's slides:

Depending upon the values of these parameters the complex exponential can exhibit

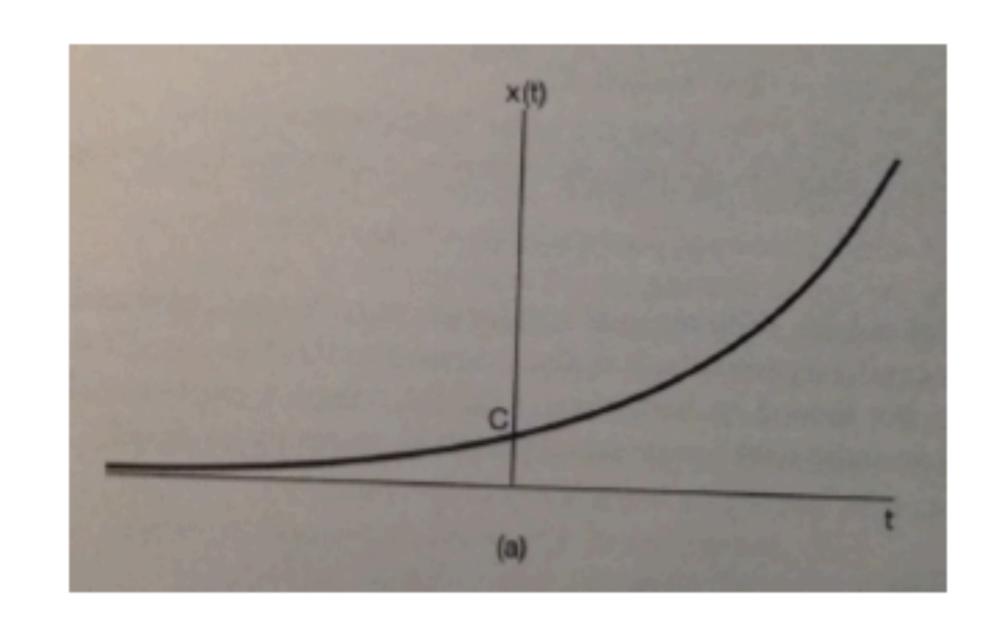
different characteristics.

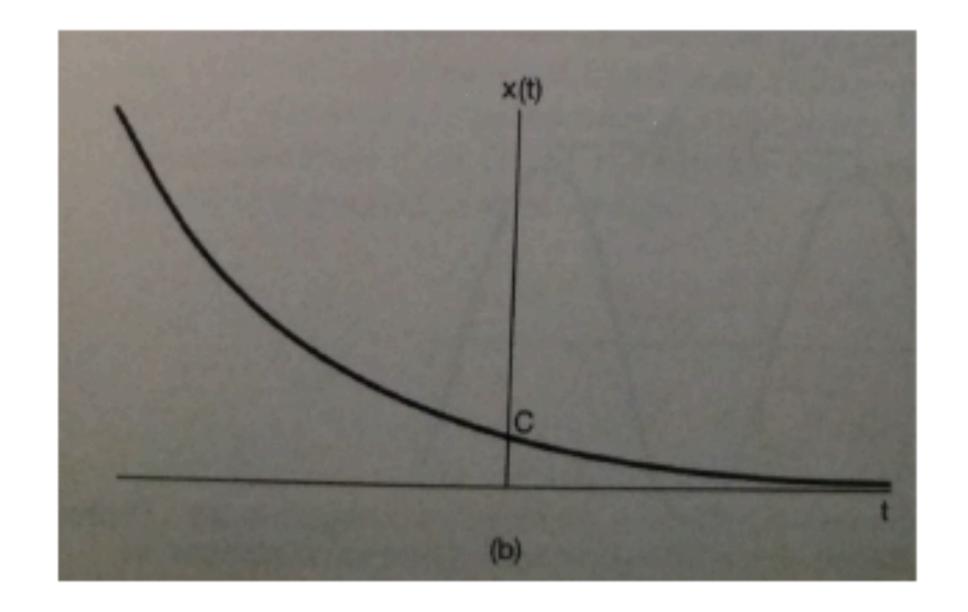


• From Laura Martínez's slides:

Real exponential:

• When $C, a \in \mathbb{R} \to x(t) = Ce^{\sigma t}$





Growing exponential($\sigma > 0$)

Decaying exponential (σ < 0)

Recall the Euler's formula:

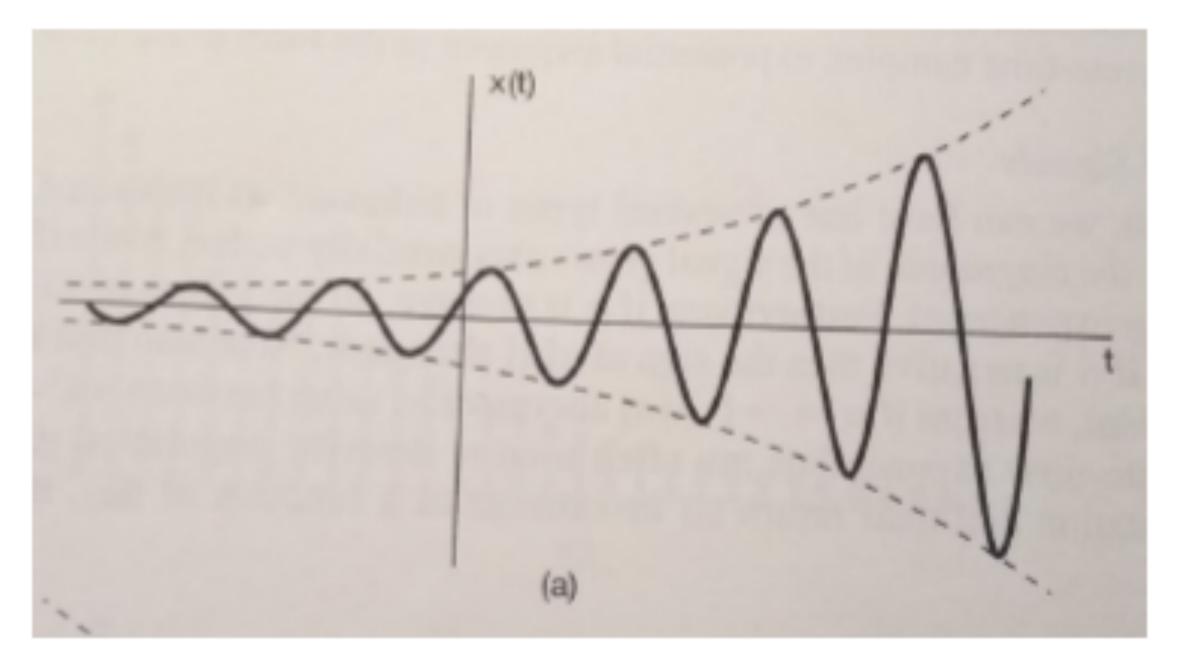
$$e^{j\omega t} = \cos(\omega t) + j\sin(\omega t)$$

$$\cos(\omega t) = \frac{e^{j\omega t} + e^{-j\omega t}}{2} \qquad \sin(\omega t) = \frac{e^{j\omega t} - e^{-j\omega t}}{2j}$$

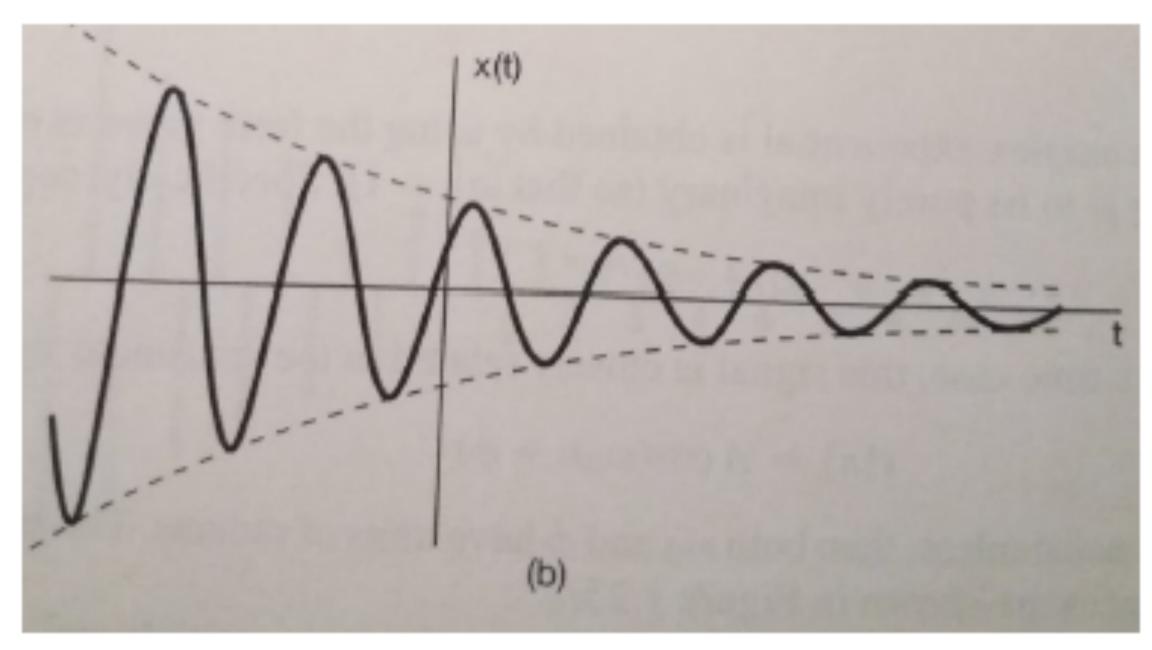
• Already linear combination of exponentials...

• From Laura Martínez's slides:

$$e^{st} = e^{\sigma t} e^{j\omega t}$$



Growing sinusoidal signal ($\sigma > 0$)



Decaying exponential (σ < 0)

Let us consider an exponential as input of an LTI system:

$$x(t) = e^{st}$$

$$y(t) = x(t) * h(t) = h(t) * x(t) = \int_{-\infty}^{+\infty} h(\tau)x(t-\tau)d\tau$$

$$= \int_{-\infty}^{+\infty} h(\tau)e^{s(t-\tau)}d\tau$$

$$= e^{st} \int_{-\infty}^{+\infty} h(\tau)e^{-s\tau} d\tau$$

Let us consider an exponential as input of an LTI system:

$$y(t) = e^{st} \int_{-\infty}^{+\infty} h(\tau)e^{-s\tau} d\tau$$

$$y(t) = H(s)e^{st}$$
$$y(t) = H(s)x(t)$$

Laplace Transform of the impulse response:

$$H(s) = \int_{-\infty}^{+\infty} h(\tau)e^{-s\tau}d\tau$$

$$y(t) = H(s)x(t)$$
$$x(t) = e^{st}$$

Laplace transformation of the impulse response:

$$H(s) = \int_{-\infty}^{+\infty} h(\tau)e^{-s\tau}d\tau$$

- Note that this integral does not always exist....it depends on:
- the function h(t) inside
- and the value of "s" (recall that is a complex number)

Laplace transformation of the a generic function/signal:

$$X(s) = \int_{-\infty}^{+\infty} x(\tau)e^{-s\tau}d\tau$$

$$X(s) = \int_{-\infty}^{+\infty} x(t)e^{-st}dt$$

 Laplace transformation is defined in the complex domain and takes complex values:

$$s \in \mathbb{C}$$
 $X(s) \in \mathbb{C}$

then usually the people plot and study:

$$|X(s)|$$
 Real $[X(s)]$ phase $[X(s)]$ Imag $[X(s)]$

Let us consider an exponential as input of an LTI system:

$$x(t) = e^{st}$$

then the output is:

$$y(t) = H(s)x(t)$$

$$|y(t)| = |H(s)||x(t)|$$

$$|y(t)| = |H(s)||x(t)|$$

• Then, when:

$$|H(s)| \to 0$$
 $|y(t)| = 0$
 $|H(s)| \to \infty$ $|y(t)| \to \infty$

- The Laplace transform can destroy the output signal,
- or makes it to diverge.... (explosion instability)

$$|H(s)| \to 0$$
 $|y(t)| = 0$
 $|H(s)| \to \infty$ $|y(t)| \to \infty$

 The Laplace transform of the impulse response h(t) says almost everything regarding the LTI system:

We want to study then

as function of s.

Example: linear combination of real exponentials and h(t)=u(t)

Example: output calculation using the system function

• Consider a LTI system characterized by h(t) = u(t). Calculate the output when the input is:

$$x(t) = Ae^{s_1t} + Be^{s_2t} + Ce^{s_3t}$$

• We start calculating the system function:

$$H(s) = \int_{-\infty}^{\infty} h(\tau)e^{-s\tau}d\tau = \int_{-\infty}^{\infty} u(\tau)e^{-s\tau}d\tau = \int_{0}^{\infty} e^{-s\tau}d\tau = _{Real(s)>0}$$
$$= \frac{1}{-s}[e^{-s\tau}]_{0}^{\infty} = \frac{1}{-s}[0-1] = \frac{1}{s}$$

Using the linearity property:

$$y(t) = H(s_1)Ae^{s_1t} + H(s_2)Be^{s_2t} + H(s_3)Ce^{s_3t} = \frac{A}{s_1}e^{s_1t} + \frac{B}{s_2}e^{s_2t}\frac{C}{s_3}e^{s_3t}$$

where we assume that $Real(s_1)$, $Real(s_2)$, $Real(s_3) > 0$.

Consider:

$$\int_{-\infty}^{+\infty} h(t)e^{-2t}dt$$

what is it: H(....) ?

$$H(2) = \int_{-\infty}^{+\infty} h(t)e^{-2t}dt$$

Consider:

$$\int_{-\infty}^{+\infty} h(t)e^{-jt}dt = ?$$

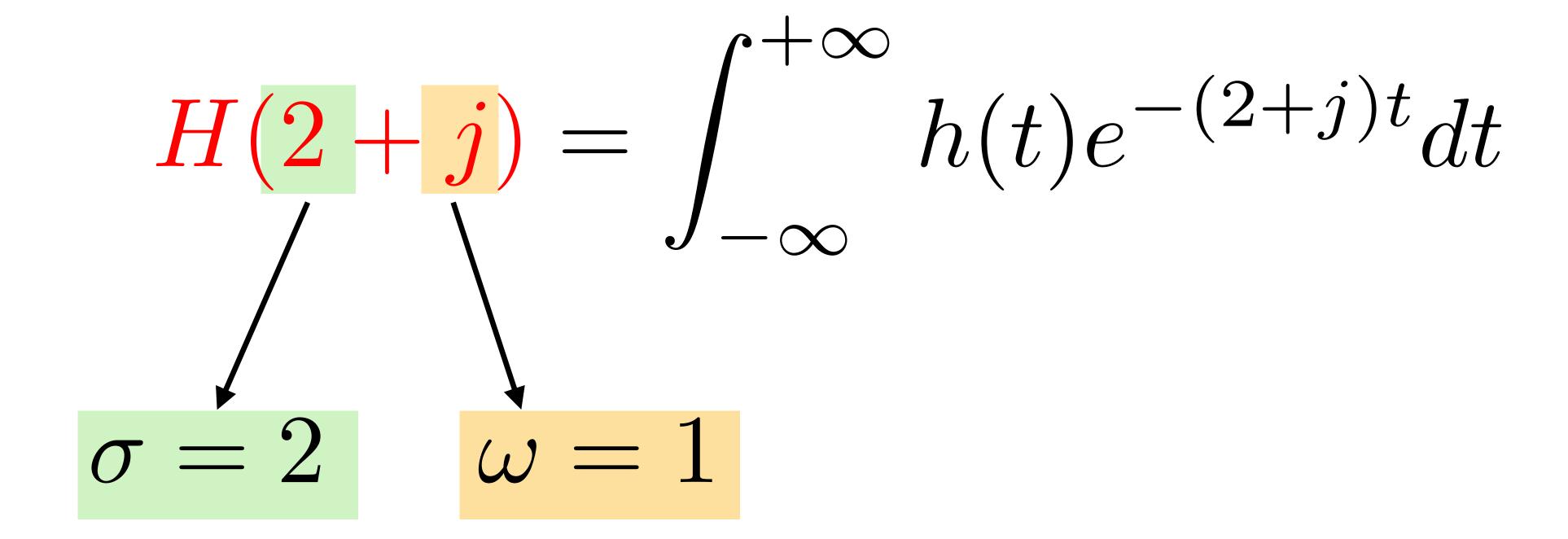
what is it: H(....) ?

$$H(j) = \int_{-\infty}^{+\infty} h(t)e^{-jt}dt$$

Consider:

$$\int_{-\infty}^{+\infty} h(t)e^{-(2+j)t}dt = ?$$

what is it: H(....) ?



STANDARD Fourier transform

Recall:

$$S = \sigma + \jmath \omega$$
• Frequency

• From Laplace Transform to Fourier Transform:

STANDARD Fourier transform

$$s = 0 + j\omega$$

• The FOURIER TRASFORM IS A SPECIAL CASE OF THE LAPLACE TRANSFORM WITH:

STANDARD Fourier transform

$$X(\omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t}dt$$

Laplace and STANDARD Fourier transforms

$$X(s) = \int_{-\infty}^{+\infty} x(t)e^{-st}dt$$
$$X(\omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t}dt$$

Laplace and STANDARD Fourier transforms

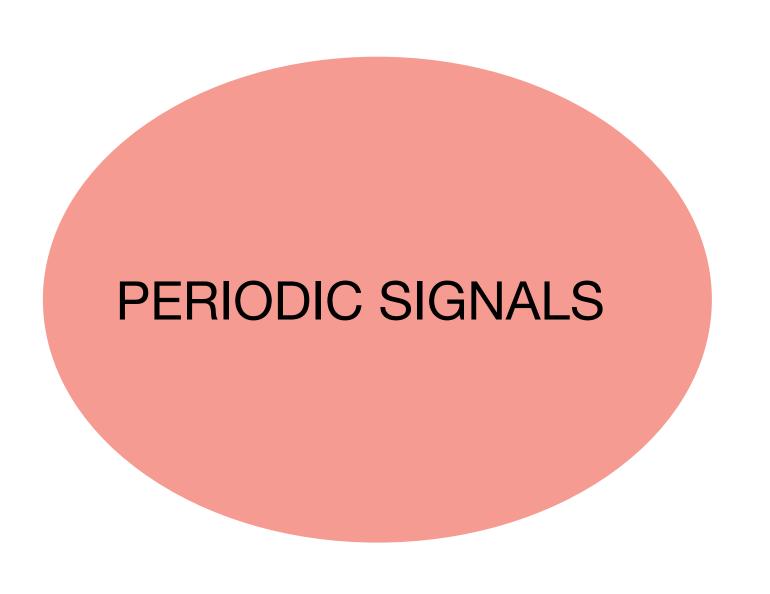
- Some signals/functions do not have Laplace Transform.
- And even less signals/functions have STANDARD Fourier Transform...

Signals with Laplace Transform

Signals with STANDARD Fourier Transform

Laplace and STANDARD Fourier transforms

• PERIODIC SIGNALS HAVE NOT LAPLACE/FOURIER



Signals with Laplace Transform

Signals with STANDARD Fourier Transform

Periodic signals in transformed domains

• PERIODIC SIGNALS HAVE NOT LAPLACE/STANDARD FOURIER.

- We will also see how to deal with periodic signals.
- For them we have the FOURIER SERIES and
- the Generalized Fourier Transform.

Questions?