

TOPIC 4

The practical computation of Fourier
(e.g., with Matlab)

PART 2

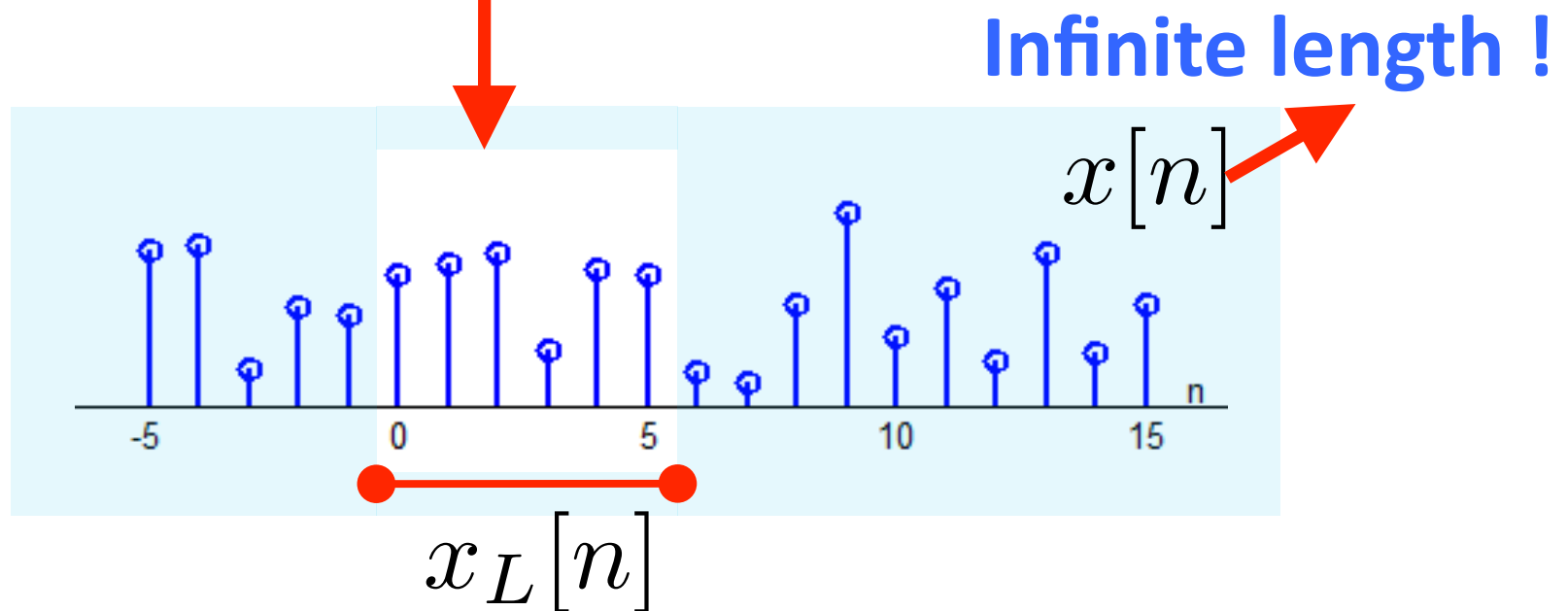
So far...

➤ We have interpreted the data (the finite sequence of values) as a periodic signal.

➤ We can consider another approach: the finite sequence of values (that we observe) is a piece of infinite signal.

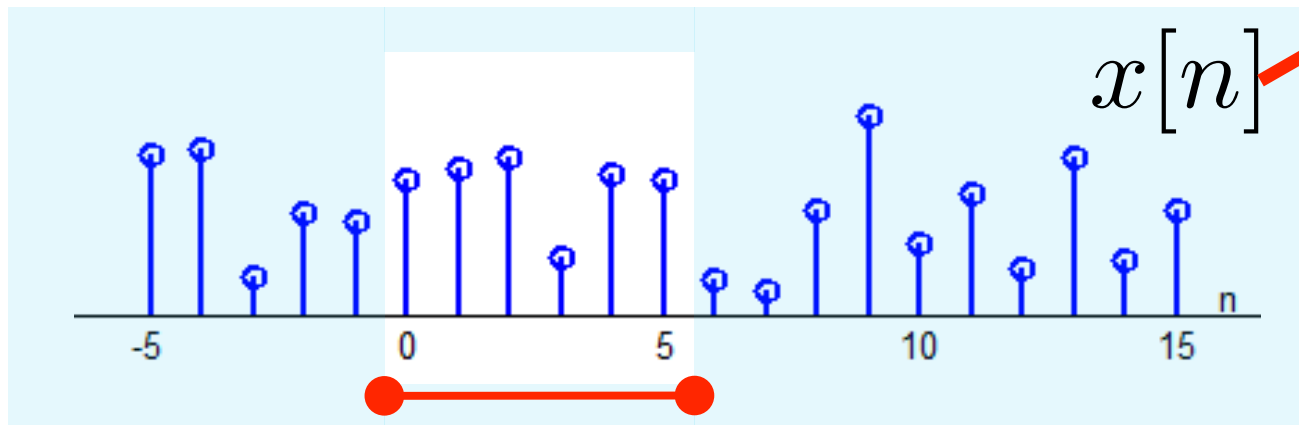
Just a “piece” of a signal...

- We can consider another approach: the finite sequence of values (that we observe) is a piece of infinite signal.



Just a “piece” of a signal...

Infinite length !



$$x_L[n] = x[n]w_L[n]$$

windowed signal

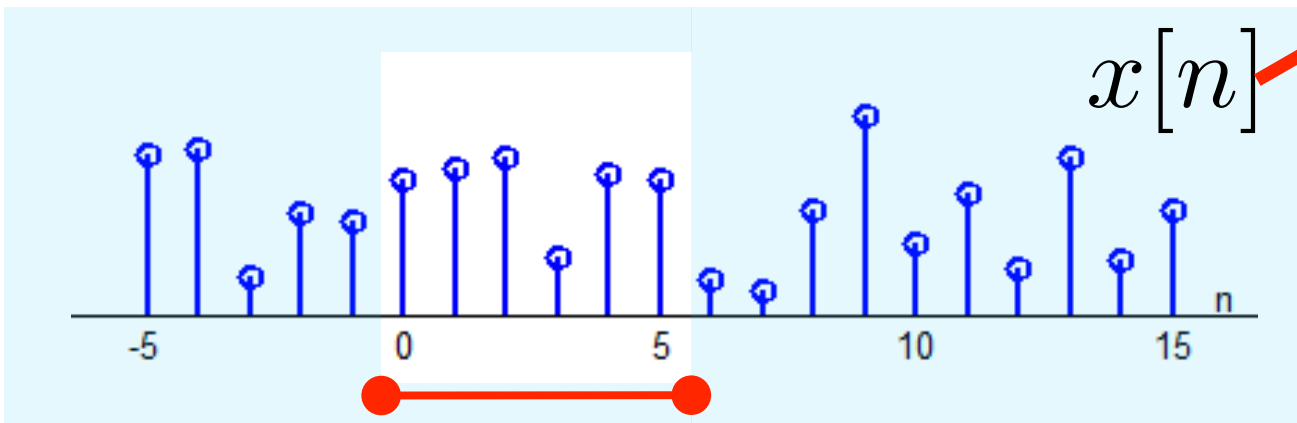
“window”

$$w_L[n] = \begin{cases} 1 & n = 0, \dots, L - 1 \\ 0 & \text{Otherwise} \end{cases}$$

Just a “piece” of a signal...

SUMMARY: Now, we interpret that we “see” a windowed signal.

Infinite length !



$$x_L[n] = x[n]w_L[n]$$

windowed signal

In frequency

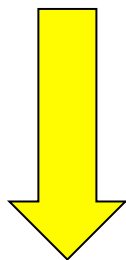
$$x_L[n] = x[n]w_L[n]$$

$$X_L(\Omega) = X(\Omega) * W_L(\Omega)$$

$$X_L(\Omega) = \frac{1}{2\pi} \int_{2\pi} X(\theta)W_L(\Omega - \theta)d\theta$$

In frequency

$$w_L[n] = \begin{cases} 1 & n = 0, \dots, L-1 \\ 0 & \text{Otherwise} \end{cases}$$

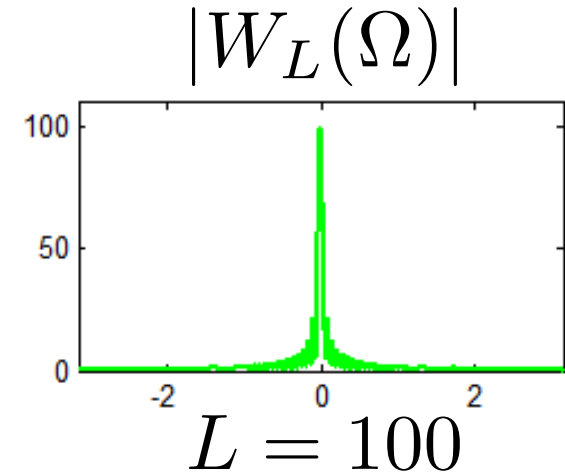
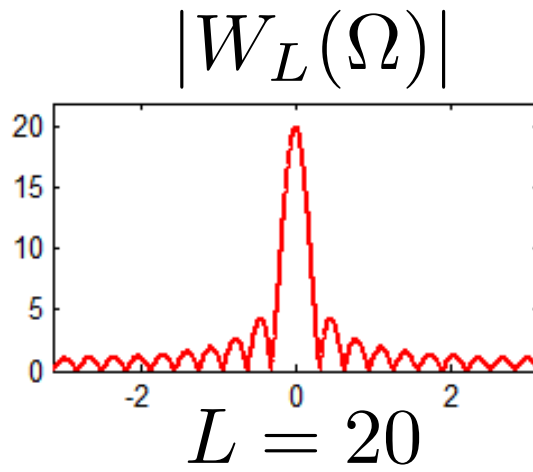
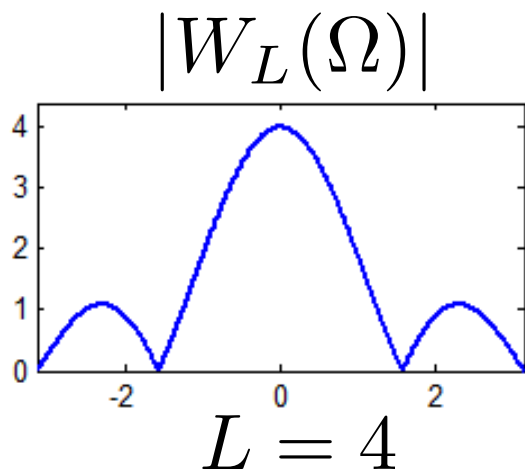


$$W_L(\Omega) = \frac{\sin\left(\frac{L\Omega}{2}\right)}{\sin\left(\frac{\Omega}{2}\right)} e^{-j\frac{(L-1)\Omega}{2}}$$

RECALL that is **PERIODIC** of period 2π

In frequency

$$W_L(\Omega) = \frac{\sin\left(\frac{L\Omega}{2}\right)}{\sin\left(\frac{\Omega}{2}\right)} e^{-j\frac{(L-1)\Omega}{2}}$$



As L grows, the FT of “the window” becomes more similar to delta in frequency.

Summary (so far)

TIME:

$$w_L[n] \xrightarrow{L \rightarrow \infty} 1$$

$$x_L[n] = x[n]w_L[n] \xrightarrow{L \rightarrow \infty} x[n]$$

Frequency:

$$W_L(\Omega) \xrightarrow{L \rightarrow \infty} 2\pi \sum_{k=-\infty}^{+\infty} \delta(\Omega - 2\pi k)$$

$$X_L(\Omega) \xrightarrow{L \rightarrow \infty} X(\Omega)$$

also

$$X_L(\Omega) = X(\Omega) * W_L(\Omega) = \frac{1}{2\pi} \int_{2\pi} X(\theta) W_L(\Omega - \theta) d\theta$$

Generalized FT

RECALL

We could always consider a larger sequence:

$$N \geq L$$

Adding zeros to the windowed signal, $x_L[n]$

(or considering the alternative mathematical view of DFT... see last slide of TOPIC 4- part 1)

Relationship with DFT

$$\begin{aligned} X_N[k] &= \sum_{n=0}^{N-1} x_L[n] e^{-jk \frac{2\pi}{N} n} \\ &= \sum_{n=0}^{L-1} x_L[n] e^{-jk \frac{2\pi}{N} n} \end{aligned}$$

$$x_L[n] = x[n] w_L[n]$$

$$X_L(\Omega) = X(\Omega) * W_L(\Omega)$$

$$X_L(\Omega) = \sum_{n=-\infty}^{+\infty} x_L[n] e^{-jk\Omega n} = \sum_{n=0}^{L-1} x_L[n] e^{-j\Omega n}$$

Relationship with DFT

COMPARING:

(Matlab gives you N numbers)

$$X_N[k] = \sum_{n=0}^{L-1} x_L[n] e^{-jk \frac{2\pi}{N} n} \quad 0 \leq k \leq N - 1$$

But actually we know that we have $X_N[k] = X_N[k + N]$

$$X_L(\Omega) = \sum_{n=0}^{L-1} x_L[n] e^{-j\Omega n} \quad -\infty < \Omega < \infty$$

Periodic of period 2π

THEN AGAIN:

$$X_N[k] = X_L(\Omega) \Big|_{k \frac{2\pi}{N}} = X_L\left(k \frac{2\pi}{N}\right)$$

Relationship with DFT

$$X_N[k] = X_L(\Omega) \Big|_{k \frac{2\pi}{N}} = X_L\left(k \frac{2\pi}{N}\right)$$

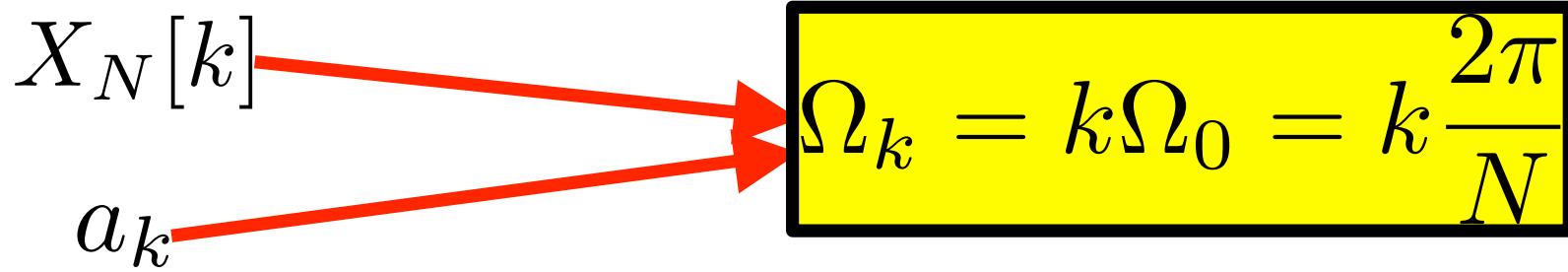
$$X_N[k] = X_L(\Omega) \Big|_{k\Omega_0} = X_L(k\Omega_0)$$

$$\Omega_0 = \frac{2\pi}{N}$$

THEN, WE ARE “SAMPLING” the FOURIER TRANSFORM $X_L(\Omega)$ in frequency domain.

Exactly the same conclusions...

Relationship with DFT



The diagram shows two mathematical expressions on the left, $X_N[k]$ and a_k , with red arrows pointing to a yellow rectangular box on the right. Inside the box is the equation $\Omega_k = k\Omega_0 = k \frac{2\pi}{N}$.

$$X_N[k]$$
$$a_k$$
$$\Omega_k = k\Omega_0 = k \frac{2\pi}{N}$$

Exactly the same conclusions of PART 1...

VERY IMPORTANT SUMMARY

(1) We have: $x[0], x[1], x[2], \dots, x[L-1]$

(2) We compute DFT (e.g., by Matlab) with


$$N \geq L$$


$$X_N[k] = \sum_{n=0}^{L-1} x_L[n] e^{-jk \frac{2\pi}{N} n}$$


(3) Matlab returns N complex numbers:

$$X_N[0], X_N[1], X_N[2], \dots, X_N[N-1]$$


$$0$$

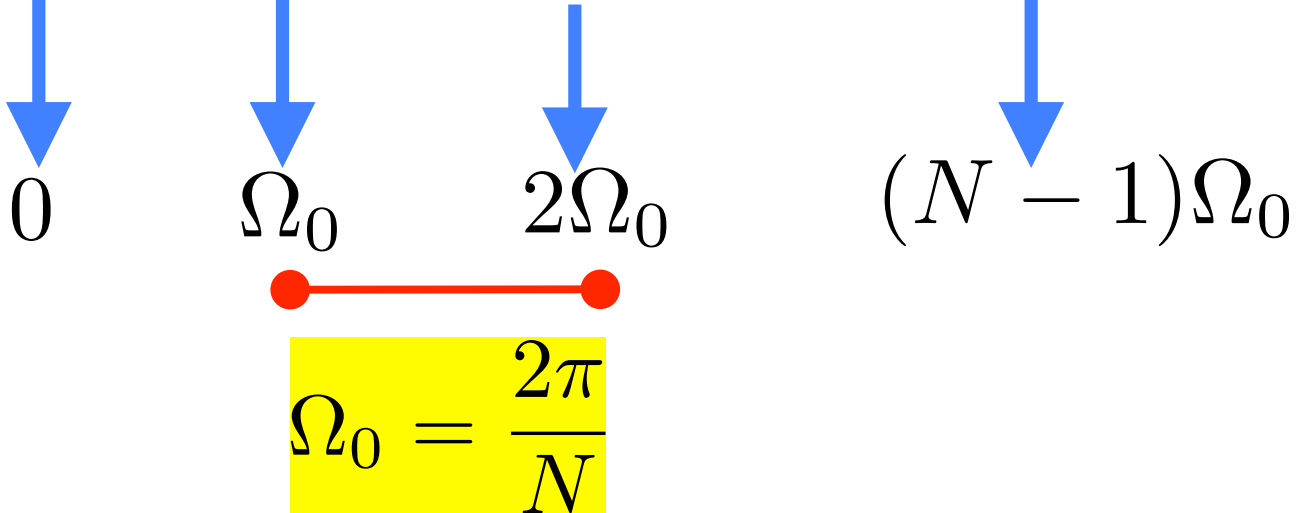

$$\Omega_0$$


$$2\Omega_0$$


$$(N-1)\Omega_0$$

VERY IMPORTANT SUMMARY

(3) Matlab returns N complex numbers:

$$X_N[0], X_N[1], X_N[2] \dots, X_N[N-1]$$


0 Ω_0 $2\Omega_0$ $(N-1)\Omega_0$

$\Omega_0 = \frac{2\pi}{N}$

$$X_N[k] = X_L(\Omega) \Big|_{k\Omega_0} = X_L(k\Omega_0)$$

VERY IMPORTANT SUMMARY

(4) Moreover the DFT values, that we get, are “samples” of the Fourier Transform of the **windowed signal**: we are assuming that we are observing just a “piece” of an infinite (non-periodic) signal...

$$X_N[k] = X_L(\Omega) \Big|_{k\Omega_0} = X_L(k\Omega_0)$$

$$x_L[n] = x[n]w_L[n]$$

$$X_L(\Omega) = X(\Omega) * W_L(\Omega)$$